

Fundamentals of Communications

Engineering

Department of Communications Engineering, College of Engineering, University of Diyala, 2016-2017

Class: Second Year

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Room: Comm-02

Lecture: 17

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Angle Modulation

Angle Modulation

Frequency Modulation (FM)

Frequency of the carrier is varied according to the message signal amplitude.

Digital format is FSK

Phase Modulation (PM)

Phase angle of the carrier is varied according to the message signal amplitude.

Digital format is PSK

* Angle Modulation used in

- ① Radio broadcasting,
- ② Two way mobile radio,
- ③ Microwave communication,
- ④ TV sound transmission,
- ⑤ Cellular radio, and
- ⑥ Satellite communication.

* Instantaneous Frequency :

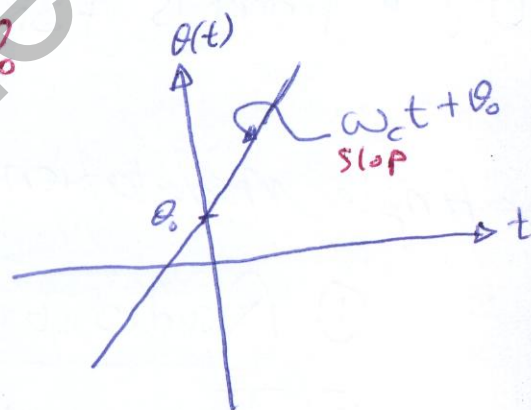
+ Consider a generalized sinusoidal signal $\phi(t)$:

$$\phi(t) = A \cos[\theta(t)] \quad \text{--- (5.1)}$$

where $\theta(t)$ is the generalized angle, and it is a function of time t .

+ The generalized angle of conventional sinusoid

$A \cos(\omega_c t + \theta_0)$ is $\omega_c t + \theta_0$



* $\omega_c t + \theta_0$ is a straight line with a slope ω_c and intercept θ_0 .

* Hence ; $\phi(t) = A \cos(\omega_c t + \theta_0)$

* The frequency of $\phi(t)$ is the slope of $\theta(t)$

* Therefore, the instantaneous frequency at any time t is ω_i

$$\omega_i(t) = \frac{d\theta}{dt} \quad \text{and}$$

(5.2a)

$$\theta(t) = \int_{-\infty}^t \omega_i(\alpha) d\alpha \quad \text{--- (5.2b)}$$

- * A modulating signal (message signal) $m(t)$ is possible to transmit it by varying the angle θ of a carrier.
- * Carrying $m(t)$ on the angle of the carrier is called angle modulation or exponential modulation.
- * Two simple possibilities are FM & PM.
- * In PM, the angle $\theta(t)$ is varied linearly with $m(t)$:-

$$\theta(t) = \omega_c t + \theta_0 + k_p m(t)$$

where k_p is a constant

and ω_c is the carrier frequency

let $\theta_0 = 0$:-

$$\theta(t) = \omega_c t + k_p m(t) \quad \text{--- (5.3a)}$$

∴ The PM wave is ∴

$$\phi_{PM}(t) = A \cos[\omega_c t + k_p m(t)] \quad \text{--- (5.3b)}$$

* The instantaneous frequency in PM thus :-

$$\omega_i(t) = \frac{d\theta}{dt} = \omega_c + k_p \frac{dm(t)}{dt}$$

$$\omega_i(t) = \omega_c + k_p \dot{m}(t) \quad \text{--- (5.3c)}$$

∴ In PM : the instantaneous frequency ω_i varies linearly with the derivative of $m(t)$

in PM $\omega_i(t) \propto \dot{m}(t)$

* If ω_i varies linearly with $m(t)$ itself not with $\dot{m}(t)$, then we have FM modulation

* In FM : ω_i varies linearly with $m(t)$

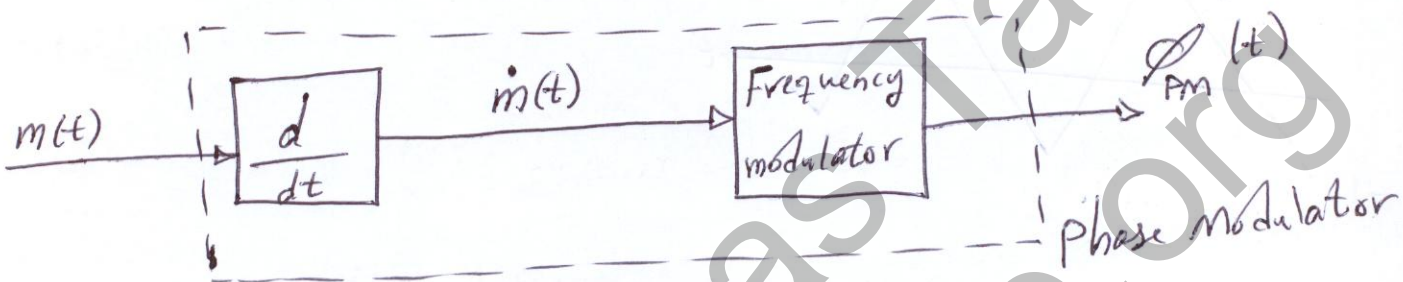
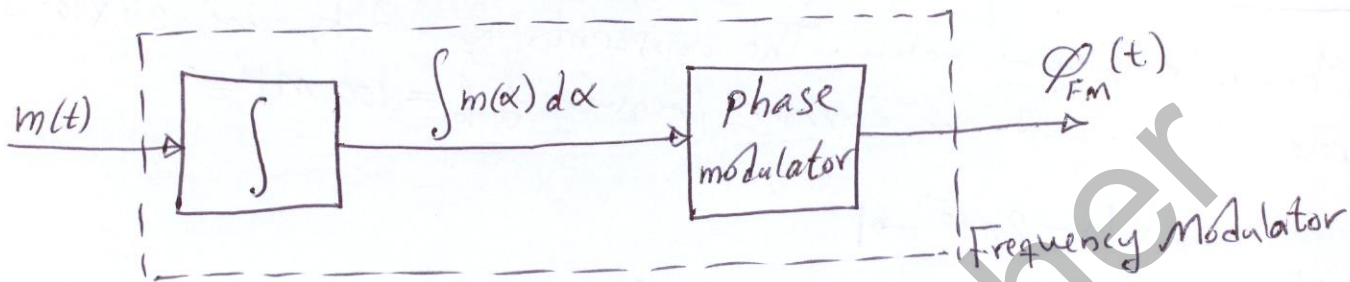
$$\omega_i(t) = \omega_c + k_f m(t) \quad \text{--- (5.4a)}$$

where k_f is a constant.

$$\theta(t) = \int_{-\infty}^t [\omega_c + k_f m(\alpha)] d\alpha = \omega_c t + k_f \int_{-\infty}^t m(\alpha) d\alpha \quad \text{--- (5.4b)}$$

∴ FM wave is :-

$$\phi_{FM}(t) = A \cos \left[\omega_c t + k_f \int_{-\infty}^t m(\alpha) d\alpha \right] \quad \text{--- (5.4c)}$$



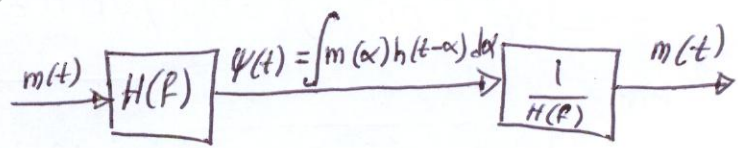
* The generalized angle-modulated carrier $\phi_{EM}(t)$ can be:-

$$\phi_{EM}(t) = A \cos(\omega_c t + \psi) \quad \text{--- (5.5 a)}$$

$$= A \cos \left[\omega_c t + \int_{-\infty}^{\infty} m(\alpha) h(t-\alpha) d\alpha \right] \quad \text{--- (5.5 b)}$$

* For PM ::

$$h(t) = k_p \delta(t)$$

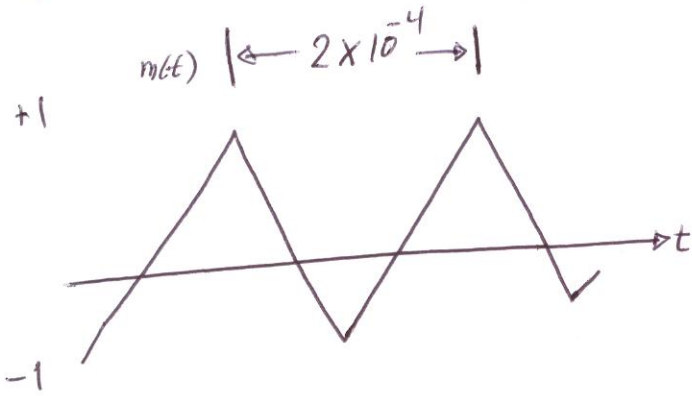


* For FM ::

$$h(t) = k_f u(t)$$

* where by this generalization $h(t)$ can be any other function. Hence, millions of angle modulations can be produced.

EX. 5.1 Sketch FM and PM waves for the modulating signal $m(t)$ shown in Figure below. The constants k_f and k_p are $2\pi \times 10^5$ and 10π , respectively, and the carrier frequency $f_c = 100$ MHz.



Solution For FM :- $\omega_i = \omega_c + k_f m(t)$

or $f_i = f_c + \frac{k_f}{2\pi} m(t) = 10^8 + 10^5 m(t)$

* minimum f_i is when $m(t)$ is minimum.

$(f_i)_{\min} = 10^8 + 10^5 [m(t)]_{\min}$

$f_{i\min} = 10^8 + 10^5 (-1) = 99.9$ MHz.

$f_{i\max} = 10^8 + 10^5 (+1) = 100.1$ MHz.

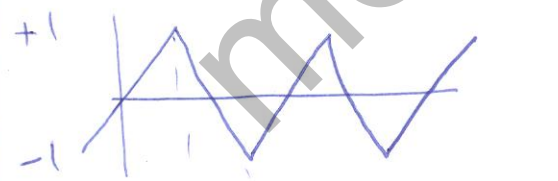
minimum $m(t) = -1$
maximum $m(t) = +1$

For PM :- $f_i = f_c + \frac{k_p}{2\pi} \dot{m}(t)$

$f_i = 10^8 + 5 \dot{m}(t)$

$f_{i\min} = 10^8 + 5 [\dot{m}(t)]_{\min} = 10^8 + 5 (-20,000) = 99.9$ MHz

$f_{i\max} = 10^8 + 5 [\dot{m}(t)]_{\max} = 10^8 + 5 (20,000) = 100.1$ MHz



Power of Angle Modulated wave :

* In angle Modulation, the frequency or the phase of the carrier varies in proportion with the amplitude of the message signal (modulating signal).

* Hence, the amplitude of the carrier is constant.

∴ Power of FM or PM modulated wave is constant :

$$P_{EM} = P_{FM} = P_{PM} = \frac{A^2}{2}$$

$$\frac{(\text{Amplitude of carrier})^2}{2}$$

Bandwidth of Angle-Modulated wave

* Let $a(t) = \int_{-\infty}^t m(x) dx$ ————— (5.6)

* Let $\hat{\phi}_{FM}(t) = A e^{j[\omega_c t + k_f a(t)]} = A e^{j k_f a(t)} e^{j \omega_c t}$ ————— (5.7a)

*→ Now the real part is

$\phi_{FM}(t) = \text{Re} \{ \hat{\phi}_{FM}(t) \}$ ————— (5.7b)

Remember the power series of e^x is

$$e^x = \sum_{n=0}^{\infty} \frac{x^n}{n!} = 1 + x + \frac{x^2}{2!} + \frac{x^3}{3!} + \dots$$

∴ $\hat{\phi}_{FM}(t) = A \left[1 + j k_f a(t) - \frac{k_f^2}{2!} a^2(t) + \dots + j^n \frac{k_f^n}{n!} a^n(t) + \dots \right] e^{j \omega_c t}$ ————— (5.8a)

*→ Now the real part is

$\phi_{FM}(t) = \text{Re} \{ \hat{\phi}_{FM}(t) \} = A \left[\underbrace{\cos \omega_c t}_{\text{carrier}} - \underbrace{k_f a(t) \sin \omega_c t}_{\text{DSB-SC}} \right]$

$\left[-\frac{k_f^2}{2!} \underbrace{a^2(t) \cos \omega_c t}_{\text{DSB-SC } a^2(t)} + \frac{k_f^3}{3!} \underbrace{a^3(t) \sin \omega_c t}_{\text{DSB-SC } a^3(t)} + \dots \right]$ ————— (5.8b)

* Since $a(t) = \text{integral of } m(t) \text{ } \therefore$

* Since $M(\omega)$ band limited to B Hz \therefore

$$a^2(t) \xleftrightarrow{\text{F.T.}} A(\omega) * A(\omega) / 2\pi$$

Bandwidth of $a^2(t) \leftrightarrow A(\omega) * \frac{A(\omega)}{2\pi}$ is $2B$ Hz.

\rightarrow Bandwidth of $a^n(t)$ is nB Hz centered around ω_c .

\therefore Theoretically, FM Bandwidth = ∞ .

Then there are two distinct possibilities in terms of bandwidth:

① narrowband FM, (NB FM)

② wideband FM. (WB FM).

* Narrow-Band Angle Modulation \therefore

* Angle modulation is non-linear unlike AM modulation.

$$* A \cos\{\omega_c t + k_f [a_1(t) + a_2(t)]\} \neq A \cos[\omega_c t + k_f a_1(t)] + A \cos[\omega_c t + k_f a_2(t)]$$

the principle of superposition does not hold. As we have studied previously.



* If $|K_f a(t)| \ll 1$, then Eq. 5.8 becomes:

$$\phi_{FM}(t) \approx A [\cos \omega_c t - K_f a(t) \sin \omega_c t] \quad (5.9)$$

* This is a linear modulation similar to AM.

* In AM, the bandwidth = $2B$ and in (5.9) $\phi_{FM}(t)$ has bandwidth = $2B$. Thus,

NBFM is when $|K_f a(t)| \ll 1$

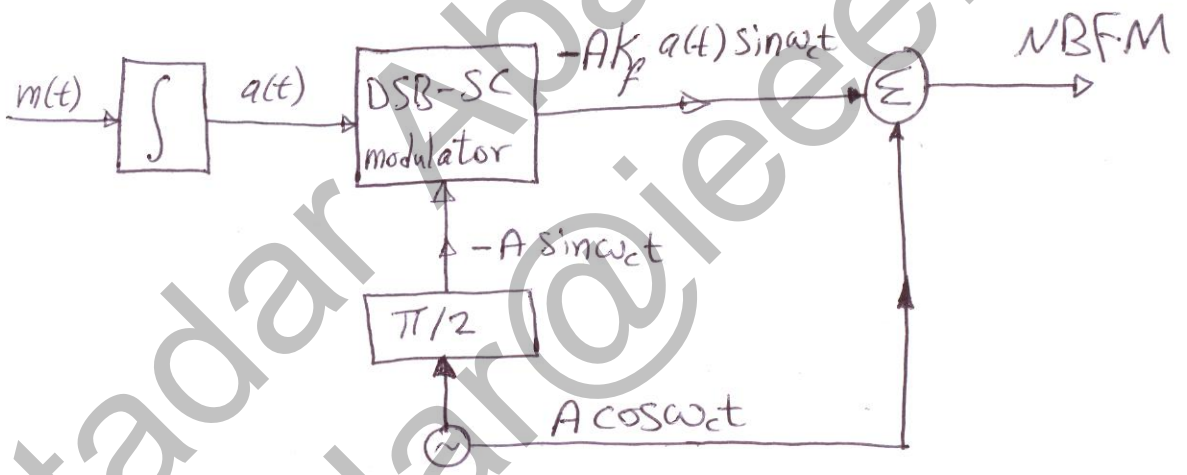
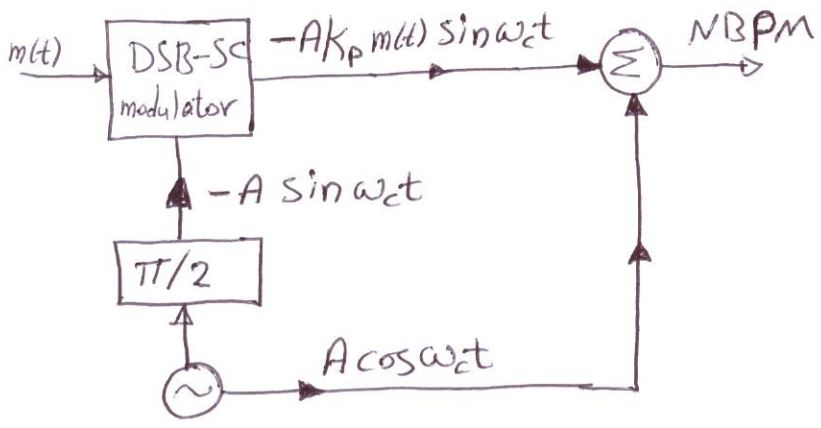
* Similarly, the PM also will be Narrow-band PM (NBFM):

$$\phi_{PM} \approx A [\cos \omega_c t - K_f m(t) \sin \omega_c t] \quad (5.10)$$

* AM and NBFM both have pure carrier and sidebands centered at $\pm \omega_c$

* Both AM & NBFM bandwidths = $2B$.

* The sideband spectrum for FM has phase shift = $\frac{\pi}{2}$ w.r.t. carrier, whereas that of AM is in phase with the carrier.



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